

# Temperature Sensor and Voltage Monitor with Simple Serial Transport

# ADT7485A

#### **FEATURES**

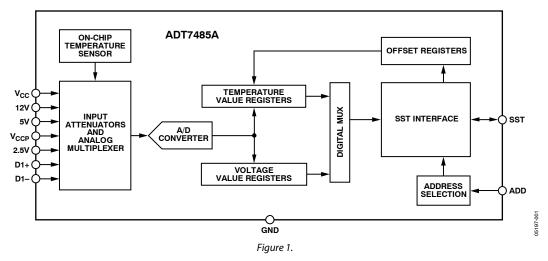
1 on-chip temperature sensor 1 remote temperature sensor Monitors up to 5 voltages Simple Serial Transport (SST) interface

#### **APPLICATIONS**

Personal computers Portable personal devices Industrial sensor nets

#### **GENERAL DESCRIPTION**

The ADT7485A is a digital temperature sensor and voltage monitor for use in PC applications with a Simple Serial Transport (SST) interface. It can monitor its own temperature as well as the temperature of a remote sensor diode. It can also monitor four external voltage channels and its own supply voltage. The ADT7485A is controlled by a single SST bidirectional data line. This device is a fixed-address SST client where the target address is chosen by the state of the address pin, ADD.



#### FUNCTIONAL BLOCK DIAGRAM

# TABLE OF CONTENTS

#### **REVISION HISTORY**

| 01/08 - Rev 2: Conversion to ON Semiconductor        |
|--|
| 12/06—Rev. 0 to Rev. A                               |
| Change to 5V Channel Input Resistance Specification  |
| Changes to Simple Serial Transport Interface Section |

7/06—Revision 0: Initial Version

| Simple Serial Transport Interface9                       |
|--|
| Voltage Measurement12                                    |
| Analog-to-Digital Converter12                            |
| Temperature Measurement14                                |
| Temperature Measurement Method14                         |
| Reading Temperature Measurements14                       |
| Simple Serial Transport Temperature Sensor Data Format15 |
| Using Discrete Transistors15                             |
| Layout Considerations15                                  |
| Temperature Offset16                                     |
| Outline Dimensions17                                     |
| Ordering Guide17   |

### **SPECIFICATIONS**

 $T_{\rm A}$  =  $T_{\rm MIN}$  to  $T_{\rm MAX},\,V_{\rm CC}$  =  $V_{\rm MIN}$  to  $V_{\rm MAX}$  unless otherwise noted.

#### Table 1.

| Parameter  | Min | Тур   | Мах     | Unit | Test Conditions/Comments   |
|--|-----|-------|---------|------|--|
| POWER SUPPLY   |     |       |         |      |  |
| Supply Voltage, Vcc  | 3.0 | 3.3   | 3.6     | V    |  |
| Undervoltage Lockout Threshold   |     | 2.8   |         | V    |  |
| Average Operating Supply Current, IDD                                  |     | 3.8   | 5       | mA   | Continuous conversions   |
| TEMPERATURE-TO-DIGITAL CONVERTER                                       |     |       |         |      |  |
| Local Sensor Accuracy  |     | +1    | ±1.75   | °C   | $40^{\circ}C \leq T_{A} \leq 70^{\circ}C; V_{CC} = 3.3 \text{ V} \pm 5\%$        |
|  |     |       | ±4      | °C   | $-40^{\circ}C \leq T_{A} \leq +100^{\circ}C$                                     |
| Remote Sensor Accuracy   |     |       | ±1      | °C   | $-40^{\circ}C \le T_{D} \le +125^{\circ}C; T_{A} = 25^{\circ}C; V_{CC} = 3.3 V$  |
|  |     | +1    | ±1.75   | °C   | $\begin{array}{ l l l l l l l l l l l l l l l l l l l$                           |
|  |     |       | ±4      | °C   | $-40^{\circ}C \le T_D \le +125^{\circ}C; -40^{\circ}C \le T_A \le +100^{\circ}C$ |
| Remote Sensor Source Current   |     | 12    | <u></u> | μA   | Low level  |
| Nemote Sensor Source Current   |     | 80    |         | μΑ   | Mid level  |
|  |     | 204   |         | μΑ   | High level   |
| Resolution   |     | 0.016 |         | °C   |  |
| Series Resistance Cancellation   |     | 1.5   |         | kΩ   | The ADT7485A cancels 1.5 kΩ in series  |
|  |     | 1.5   |         | N12  | with the remote thermal diode  |
| DIGITAL INPUT (ADD)  |     |       |         |      |  |
| Input High Voltage, V <sub>IH</sub>                                    | 2.3 |       |         | V    |  |
| Input Low Voltage, V <sub>IL</sub>                                     |     |       | 0.8     | V    |  |
| Input High Current, I <sub>IH</sub>                                    | -1  |       |         | μΑ   | $V_{IN} = V_{CC}$  |
| Input Low Current, I⊾  |     |       | 1       | μA   | $V_{IN} = 0$   |
| Pin Capacitance  |     | 5     |         | pF   |  |
| ANALOG-TO-DIGITAL CONVERTER<br>(Including Multiplexer and Attenuators) |     |       |         |      |  |
| Total Unadjusted Error (TUE)   |     |       | ±2      | %    | 12V and 5V channels  |
|  |     |       | ±1.5    | %    | For all other channels   |
| Differential Nonlinearity (DNL)  |     |       | ±1      | LSB  | 10 bits  |
| Power Supply Sensitivity   |     | ±0.1  |         | %/V  |  |
| Conversion Time (Voltage Input) <sup>1</sup>                           |     |       | 11      | ms   | Averaging enabled  |
| Conversion Time (Local Temperature) <sup>1</sup>                       |     |       | 12      | ms   | Averaging enabled  |
| Conversion Time (Remote Temperature) <sup>1</sup>                      |     |       | 38      | ms   | Averaging enabled  |
| Total Monitoring Cycle Time <sup>1</sup>                               |     | 145   |         | ms   | Averaging enabled  |
| Input Resistances  |     |       |         |      |  |
| V <sub>CCP</sub> and 2.5V Channels                                     | 80  | 110   | 140     | kΩ   |  |
| 5V Channel   | 95  | 120   | 150     | kΩ   |  |
| 12V Channel  | 180 | 230   | 280     | kΩ   |  |
| DIGITAL I/O (SST Pin)  |     |       |         |      |  |
| Input High Voltage , V <sub>IH</sub>                                   | 1.1 |       |         | V    |  |
| Input Low Voltage, V <sub>L</sub>                                      |     |       | 0.4     | V    |  |
| Hysteresis <sup>1</sup>  |     | 150   |         | mV   | Between input switching levels   |
| Output High Voltage, Vон   | 1.1 |       | 1.9     | V    | $I_{SOURCE} = 6 \text{ mA} (\text{maximum})$                                     |
| High Impedance State Leakage, ILEAK                                    |     |       | ±1      | μΑ   | Device powered on SST bus;<br>V <sub>SST</sub> = 1.1 V, V <sub>CC</sub> = 3.3 V  |
| High Impedance State Leakage, ILEAK                                    |     |       | ±10     | μΑ   | Device unpowered on SST bus;<br>$V_{SST} = 1.1 V$ , $V_{CC} = 0 V$               |

| Parameter  | Min                        | Тур                         | Max                          | Unit   | Test Conditions/Comments  |
|--|----------------------------|-----------------------------|------------------------------|--------|---|
| Signal Noise Immunity, V <sub>NOISE</sub>            | 300                        |                             |                              | mV p-p | Noise glitches from 10 MHz to 100 MHz;<br>width up to 50 ns             |
| SST TIMING   |                            |                             |                              |        |   |
| Bitwise Period, tBIT                                 | 0.495                      |                             | 500                          | μs     |   |
| High Level Time for Logic 1, $t_{\text{H1}}^2$       | $0.6 	imes t_{\text{BIT}}$ | $0.75 	imes t_{\text{BIT}}$ | $0.8 	imes t_{\text{BIT}}$   | μs     | tBIT defined in speed negotiation                                       |
| High Level Time for Logic 0, $t_{H0}^2$              | $0.2 	imes t_{\text{BIT}}$ | $0.25 	imes t_{\text{BIT}}$ | $0.4 	imes t_{\text{BIT}}$   | μs     |   |
| Time to Assert SST High for Logic 1, tsu, нібн       |                            |                             | $0.2 	imes t_{\text{BIT}}$   | μs     |   |
| Hold Time, t <sub>HOLD</sub> <sup>3</sup>            |                            |                             | $0.5 	imes t_{\text{BIT-M}}$ | μs     | See Simple Serial Transport<br>Specification Rev 1.0                    |
| Stop Time, t <sub>STOP</sub>                         | $1.25 	imes t_{BIT}$       | $2 \times t_{\text{BIT}}$   | $2 \times t_{\text{BIT}}$    | μs     | Device responding to a constant low level driven by originator          |
| Time to Respond After a Reset, t <sub>RESET</sub>    |                            |                             | 0.4                          | ms     |   |
| Response Time to Speed Negotiation<br>After Power-Up |                            | 500                         |                              | μs     | Time after power-up when device can<br>participate in speed negotiation |

<sup>1</sup> Guaranteed by design, not production tested. <sup>2</sup> Minimum and maximum bit times are relative to t<sub>BIT</sub> defined in the timing negotiation pulse. <sup>3</sup> Device is compatible with hold time specification as driven by SST originator.

### **ABSOLUTE MAXIMUM RATINGS**

#### Table 2.

| 1 ubic 2.                                    |                  |
|--|------------------|
| Parameter                                    | Rating           |
| Supply Voltage (V <sub>CC</sub> )            | 4 V              |
| Voltage on 12V Pin                           | 16 V             |
| Voltage on 5V Pin                            | 7 V              |
| Voltage on 2.5V and $V_{CCP}$ Pins           | 3.6 V            |
| Voltage on Any Other Pin (Including SST Pin) | –0.3 V to +3.6 V |
| Input Current at Any Pin                     | ±5 mA            |
| Package Input Current                        | ±20 mA           |
| Maximum Junction Temperature (TJ max)        | 150°C            |
| Storage Temperature Range                    | –65°C to +150°C  |
| Lead Temperature, Soldering                  |                  |
| IR Peak Reflow Temperature                   | 260°C            |
| Lead Temperature (10 sec)                    | 300°C            |
| ESD Rating                                   | 1500 V           |

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### THERMAL RESISTANCE

 $\theta_{JA}$  is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

#### Table 3. Thermal Resistance

| Package Type | Αιθ | ονθ | Unit |
|--------------|-----|-----|------|
| 10-Lead MSOP | 206 | 44  | °C/W |

#### ESD CAUTION



**ESD** (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## **PIN CONFIGURATION AND FUNCTIONAL DESCRIPTIONS**



#### **Table 4. Pin Function Descriptions**

| Pin No. | Mnemonic         | Туре                 | Description  |
|---------|------------------|----------------------|--|
| 1       | Vcc              | Power supply         | 3.3 V $\pm$ 10%. V <sub>CC</sub> is also monitored through this pin. |
| 2       | GND              | Ground               | Ground Pin.  |
| 3       | D1+              | Analog input         | Positive Connection to Remote 1 Temperature Sensor.                  |
| 4       | D1–              | Analog input         | Negative Connection to Remote 1 Temperature Sensor.                  |
| 5       | 12V              | Analog input         | 12 V Supply Monitor.   |
| 6       | 5V               | Analog input         | 5 V Supply Monitor.  |
| 7       | V <sub>CCP</sub> | Analog input         | Processor Core Voltage Monitor.                                      |
| 8       | 2.5V             | Analog input         | 2.5 V Supply Monitor.  |
| 9       | ADD              | Digital input        | Simple Serial Transport (SST) Address Select.                        |
| 10      | SST              | Digital input/output | Simple Serial Transport (SST) Bidirectional Data Line.               |

### **TYPICAL PERFORMANCE CHARACTERISTICS**

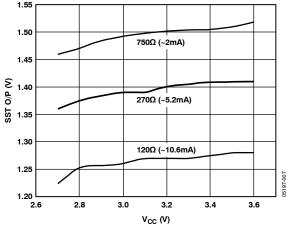


Figure 3. SST O/P Level vs. Supply Voltage

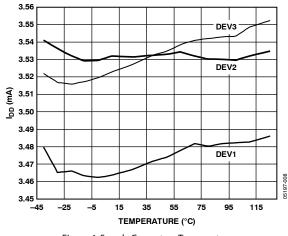
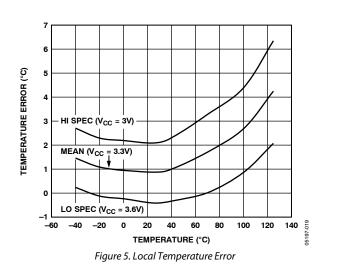


Figure 4. Supply Current vs. Temperature



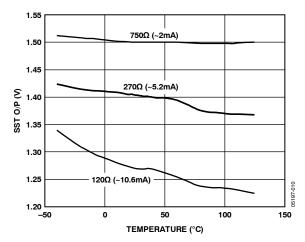
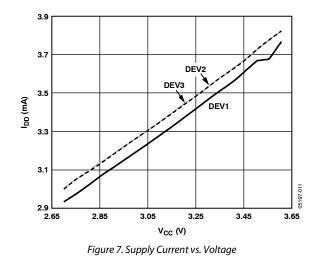
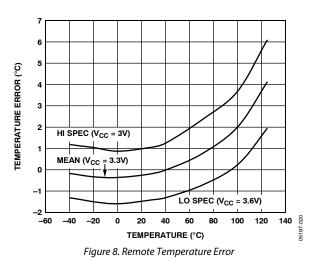
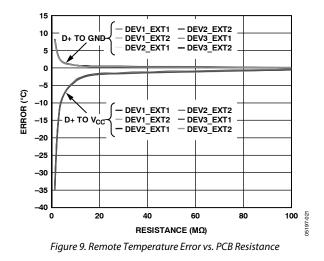


Figure 6. SST O/P Level vs. Temperature





Rev. 2 | Page 7 of 17 | www.onsemi.com



30 25 TEMPERATURE ERROR (°C) 20 100mV 15 60mV 10 5 40m\ 0 10k 100k 1M 10M 100M 1G -014 6351 NOISE FREQUENCY (Hz)

Figure 10. Temperature Error vs. Common-Mode Noise Frequency

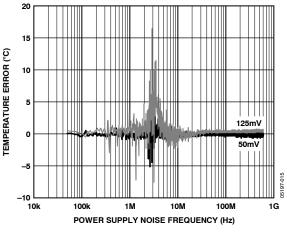


Figure 11. Local Temperature Error vs. Power Supply Noise

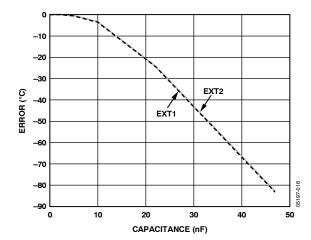


Figure 12. Remote Temperature Error vs. Capacitance Between D1+ and D1-

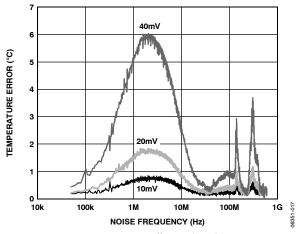


Figure 13. Temperature Error vs. Differential-Mode Noise Frequency

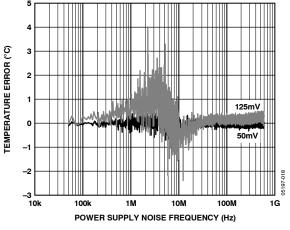


Figure 14. Remote Temperature Error vs. Power Supply Noise

### **PRODUCT DESCRIPTION**

The ADT7485A is a temperature- and voltage-monitoring device. The ADT7485A can monitor the temperature of one remote sensor diode, plus its own internal temperature. It can also monitor up to five voltage channels, including its own supply voltage.

#### SIMPLE SERIAL TRANSPORT INTERFACE

Simple Serial Transport (SST) is a one-wire serial bus and a communications protocol between components intended for use in personal computers, personal hand-held devices, or other industrial sensor nets. The ADT7485A supports Simple Serial Transport Specification Rev 1.0.

Simple Serial Transport is a licensable bus technology from Analog Devices, Inc., and Intel Corporation. To inquire about obtaining a copy of the Simple Serial Transport Specification or an SST technology license, please email Analog Devices at sst\_licensing@analog.com or write to Analog Devices, 3550 North First Street, San Jose, CA 95134, Attention: SST Licensing, M/S B7-24.

#### ADT7485A Client Address

The client address for the ADT7485A is selected using the address pin. The address pin is connected to a float detection circuit, which allows the ADT7485A to distinguish between three input states: high, low (GND), and floating. The address range for the fixed address, discoverable device is 0x48 to 0x4A.

#### Table 5. ADT7485A Selectable Addresses

| ADD       | Address Selected |
|-----------|------------------|
| Low (GND) | 0x48             |
| Float     | 0x49             |
| High      | 0x4A             |

#### **Command Summary**

Table 6 summarizes the commands supported by the ADT7485A device when directed at the target address selected by the fixed address pin. It contains the command name, command code (CC), write data length (WL), read data length (RL), and a brief description.

| Command        | Command Code, CC | Write Length, WL | Read Length, RL | Description  |
|----------------|------------------|------------------|-----------------|--|
| Ping()         | 0x00             | 0x00             | 0x00            | Shows a nonzero FCS over the header if present.  |
| GetIntTemp()   | 0x00             | 0x01             | 0x02            | Shows the temperature of the device's internal thermal diode.  |
| GetExtTemp()   | 0x01             | 0x01             | 0x02            | Shows the temperature of the external thermal diode.   |
| GetAllTemps()  | 0x00             | 0x01             | 0x04            | Shows a 4-byte block of data (GetIntTemp, GetExtTemp).   |
| GetVolt12V()   | 0x10             | 0x01             | 0x02            | Shows the voltage attached to 12V input.   |
| GetVolt5V()    | 0x11             | 0x01             | 0x02            | Shows the voltage attached to 5V input.  |
| GetVoltVCC()   | 0x12             | 0x01             | 0x02            | Shows the voltage attached to V <sub>cc</sub> input.   |
| GetVolt2.5V()  | 0x13             | 0x01             | 0x01            | Shows the voltage attached to 2.5V input.  |
| GetVoltVCCP()  | 0x14             | 0x01             | 0x02            | Shows the voltage attached to V <sub>CCP</sub> input.  |
| GetAllVolts()  | 0x10             | 0x01             | 0x10            | Shows all voltage measurement values.  |
| SetExtOffset() | 0xe0             | 0x02             | 0x00            | Sets the offset used to correct errors in the external diode.  |
| GetExtOffset() | 0xe0             | 0x01             | 0x01            | Shows the offset that the device is using to correct errors in the external diode.                     |
| ResetDevice()  | 0xf6             | 0x01             | 0x00            | Functional reset. The ADT7485A also responds to this command when directed to the Target Address 0x00. |
| GetDIB()       | 0xf7             | 0x01             | 0x08            | Shows information used by SW to identify the device's  |
|                | 0xf7             | 0x01             | 0x10            | capabilities. Can be in 8- or 16-byte format.  |

Table 6. Command Code Summary

#### **Command Code Details**

#### ADT7485A Device Identifier Block

The GetDIB() command retrieves the device identifier block (DIB), which provides information to identify the capabilities of the ADT7485A. The data returned can be in 8- or 16-byte format. The full 16 bytes of DIB is detailed in Table 7. The 8-byte format involves the first eight bytes described in this table. Byte-sized data is returned in the respective fields as it appears in Table 7. Word-sized data, including vendor ID, device ID, and data values use little endian format, that is, the LSB is returned first, followed by the MSB.

| Byte | Name                     | Value           | Description  |
|------|--------------------------|-----------------|--|
| 0    | Device Capabilities      | 0xc0            | Fixed address device                                     |
| 1    | Version/Revision         | 0x10            | Meets Version 1 of the<br>SST specification              |
| 2, 3 | Vendor ID                | 00x11d4         | Contains company ID<br>number in little endian<br>format |
| 4, 5 | Device ID                | 0x7485          | Contains device ID<br>number in little<br>endian format  |
| 6    | Device Interface         | 0x01            | SST device   |
| 7    | Function Interface       | 0x00            | Reserved   |
| 8    | Reserved                 | 0x00            | Reserved   |
| 9    | Reserved                 | 0x00            | Reserved   |
| 10   | Reserved                 | 0x00            | Reserved   |
| 11   | Reserved                 | 0x00            | Reserved   |
| 12   | Reserved                 | 0x00            | Reserved   |
| 13   | Reserved                 | 0x00            | Reserved   |
| 14   | Revision ID              | 0x05            | Contains revision ID                                     |
| 15   | Client Device<br>Address | 0x48 to<br>0x4a | Dependent on the state of the address pin                |

#### Table 7. 16-Byte DIB Details

#### Ping()

The Ping() command verifies if a device is responding at a particular address. The ADT7485A shows a valid nonzero FCS in response to the Ping() command when correctly addressed.

| Table 0, Thig() Command | Table | 8. Ping() | Command |
|-------------------------|-------|-----------|---------|
|-------------------------|-------|-----------|---------|

| Target Address  | Target Address Write Length |      | FCS |
|-----------------|-----------------------------|------|-----|
| (Not necessary) | 0x00                        | 0x00 |     |

#### ResetDevice()

This command resets the register map and conversion controller. The reset command can be global or directed at the client address of the ADT7485A.

#### Table 9. ResetDevice() Command

| Target Address | Write<br>Length | Read<br>Length | Reset<br>Command | FCS |
|----------------|-----------------|----------------|------------------|-----|
| Device Address | 0x01            | 0x00           | 0xf6             |     |

#### GetIntTemp()

The ADT7485A shows the local temperature of the device in response to the GetIntTemp() command. The data has a little endian, 16-bit, twos complement format.

#### GetExtTemp()

Prompted by the GetExtTemp() command, the ADT7485A shows the temperature of the remote diode in little endian, 16-bit, twos complement format. The ADT7485A shows 0x8000 in response to this command if the external diode is an open or short circuit.

#### GetAllTemps()

The ADT7485A shows the local and remote temperatures in a 4-byte block of data (internal temperature first, followed by external temperature) in response to a GetAllTemps() command.

#### SetExtOffset()

This command sets the offset that the ADT7485A will use to correct errors in the external diode. The offset is set in little endian, 16-bit, twos complement format. The maximum offset is  $\pm 128^{\circ}$ C with  $\pm 0.25^{\circ}$ C resolution.

#### GetExtOffset()

This command causes the ADT7485A to show the offset that it is using to correct errors in the external diode. The offset value is returned in little endian format, that is, LSB before MSB.

#### ADT7485A Response to Unsupported Commands

A full list of command codes supported by the ADT7485A is given in Table 6. The offset registers (Command Code 0xe0) are the only registers that the user can write to. The other defined registers are read only. Writing to Register Addresses 0x02, 0x09, and 0x15 to 0xdf shows a valid FSC, but no action is taken by the ADT7485A. The ADT7485A shows an invalid FSC if the user attempts to write to the device between Command Codes 0xe2 to 0xee. These registers are reserved for the manufacturer's use only, and no data can be written to the device via these addresses.

## **VOLTAGE MEASUREMENT**

The ADT7485A has four external voltage measurement channels. It can also measure its own supply voltage,  $V_{CC}$ . Pin 5 to Pin 8 can measure the supplies of the 12V, 5V, processor core voltage ( $V_{CCP}$ ), and 2.5V pins, respectively. The  $V_{CC}$  supply voltage measurement is carried out through the  $V_{CC}$  pin (Pin 1). The 2.5V pin can be used to monitor a chipset supply voltage in a computer system.

#### ANALOG-TO-DIGITAL CONVERTER

All analog inputs are multiplexed into the on-chip, successive approximation, analog-to-digital converter (ADC). This has a resolution of 10 bits. The basic input range is 0 V to 2.25 V, but the inputs have built-in attenuators to allow measurement of 2.5 V, 3.3 V, 5 V, 12 V, and the processor core voltage ( $V_{CCP}$ ) without any external components.

To allow for the tolerance of these supply voltages, the ADC produces a specific output for each nominal input voltage and therefore has adequate headroom to cope with overvoltages. The full-scale voltage that can be recorded for each channel is shown in Table 10.

#### Table 10. Maximum Reported Input Voltages

| Voltage Channel | Full-Scale Voltage |
|-----------------|--------------------|
| 12V             | 16 V               |
| 5V              | 8 V                |
| Vcc             | 4 V                |
| 2.5V            | 4 V                |
| Vccp            | 4 V                |

#### Input Circuitry

The internal structure for the analog inputs is shown in Figure 15. The input circuit consists of an input protection diode and an attenuator, plus a capacitor that forms a firstorder, low-pass filter to provide input immunity to high frequency noise.

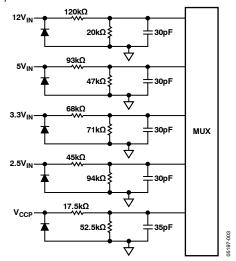


Figure 15. Internal Structure of Analog Inputs

#### Voltage Measurement Command Codes

The voltage measurement command codes are detailed in Table 11. Each voltage measurement has a read length of two bytes in little endian format (LSB followed by MSB). All voltages can be read together by addressing Command Code 0x10 with a read length of 0x10. The data is retrieved in the order listed in Table 11.

Table 11. Voltage Measurement Command Codes

| Voltage Channel  | Command Code | Returned Data |
|------------------|--------------|---------------|
| 12V              | 0x10         | LSB, MSB      |
| 5V               | 0x11         | LSB, MSB      |
| Vcc              | 0x12         | LSB, MSB      |
| 2.5V             | 0x13         | LSB, MSB      |
| V <sub>CCP</sub> | 0x14         | LSB, MSB      |

#### Voltage Data Format

The returned voltage value is in twos complement, 16-bit, binary format. The format is structured so that voltages in the range of  $\pm 32$  V can be reported. In this way, the reported value represents the number of 1/1024 V in the actual reading, allowing a resolution of approximately 1 mV.

|         | Twos Co   | Twos Complement |  |
|---------|-----------|-----------------|--|
| Voltage | MSB       | LSB             |  |
| 12 V    | 0011 0000 | 0000 0000       |  |
| 5 V     | 0001 0100 | 0000 0000       |  |
| 3.3 V   | 0000 1101 | 0011 0011       |  |
| 3 V     | 0000 1100 | 0000 0000       |  |
| 2.5 V   | 0000 1010 | 0000 0000       |  |
| 1 V     | 0000 0100 | 0000 0000       |  |
| 0 V     | 0000 0000 | 0000 0000       |  |

### **TEMPERATURE MEASUREMENT**

The ADT7485A has two dedicated temperature measurement channels: one for measuring the temperature of an on-chip band gap temperature sensor, and one for measuring the temperature of a remote diode, usually located in the CPU or GPU.

The ADT7485A monitors one local and one remote temperature channel. Monitoring of each of the channels is done in a roundrobin sequence. The monitoring sequence is in the order shown in Table 13.

#### Table 13. Temperature Monitoring Sequence

| Channel<br>Number | Measurement        | Conversion Time (ms) |
|-------------------|--------------------|----------------------|
| 0                 | Local temperature  | 12                   |
| 1                 | Remote temperature | 38                   |

#### **TEMPERATURE MEASUREMENT METHOD**

A simple method for measuring temperature is to exploit the negative temperature coefficient of a diode by measuring the base-emitter voltage ( $V_{BE}$ ) of a transistor operated at constant current. Unfortunately, this technique requires calibration to null the effect of the absolute value of  $V_{BE}$ , which varies from device to device.

The technique used in the ADT7485A measures the change in  $V_{\text{BE}}$  when the device is operated at three different currents.

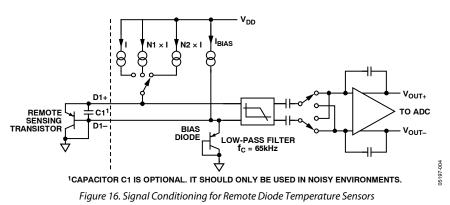
Figure 16 shows the input signal conditioning used to measure the output of a remote temperature sensor. This figure shows the remote sensor as a substrate transistor, which is provided for temperature monitoring on some microprocessors, but it could also be a discrete transistor. If a discrete transistor is used, the collector is not grounded and should be linked to the base. To prevent ground noise from interfering with the measurement, the more negative terminal of the sensor is not referenced to ground, but is biased above ground by an internal diode at the D1– input. If the sensor is operating in an extremely noisy environment, C1 can be added as a noise filter. Its value should not exceed 1000 pF. To measure  $\Delta V_{BE}$ , the operating current through the sensor is switched between three related currents. Figure 16 shows N1  $\times$  I and N2 × I as different multiples of the current I. The currents through the temperature diode are switched between I and N1 × I, giving  $\Delta V_{BE1}$ , and then between I and N2 × I, giving  $\Delta V_{BE2}$ . The temperature can then be calculated using the two  $\Delta V_{BE}$  measurements. This method can also cancel the effect of series resistance on the temperature measurement. The resulting  $\Delta V_{BE}$  waveforms are passed through a 65 kHz low-pass filter to remove noise and then through a chopper-stabilized amplifier to amplify and rectify the waveform, producing a dc voltage proportional to  $\Delta V_{BE}$ . The ADC digitizes this voltage, and a temperature measurement is produced. To reduce the effects of noise, digital filtering is performed by averaging the results of 16 measurement cycles for low conversion rates. Signal conditioning and measurement of the internal temperature sensor is performed in the same manner.

#### **READING TEMPERATURE MEASUREMENTS**

The temperature data returned is two bytes in little endian format, that is, LSB before MSB. All temperatures can be read together by using Command Code 0x00 with a read length of 0x04. The command codes and returned data are described in Table 14.

#### Table 14. Temperature Channel Command Codes

| Temp Channel | Command Code Returned Data |                             |
|--------------|----------------------------|-----------------------------|
| Internal     | 0x00                       | LSB, MSB                    |
| External     | 0x01                       | LSB, MSB                    |
| All Temps    | 0x00                       | Internal LSB, Internal MSB; |
|              |                            | External LSB, External MSB  |



# SIMPLE SERIAL TRANSPORT TEMPERATURE SENSOR DATA FORMAT

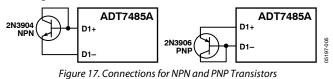
The data for temperature is structured to allow values in the range of  $\pm 512^{\circ}$ C to be reported. Thus, the temperature sensor format uses a twos complement, 16-bit binary value to represent values in this range. This format allows temperatures to be represented with approximately a 0.016°C resolution.

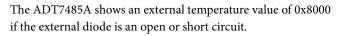
|                  | Twos Complement |           |
|------------------|-----------------|-----------|
| Temperature (°C) | MSB             | LSB       |
| -125             | 1110 0000       | 1100 0000 |
| -80              | 1110 1100       | 0000 0000 |
| -40              | 1111 0110       | 0000 0000 |
| -20              | 1111 1011       | 0011 1110 |
| -5               | 1111 1110       | 1100 0000 |
| -1               | 1111 1111       | 1100 0000 |
| 0                | 0000 0000       | 0000 0000 |
| +1               | 0000 0000       | 0100 0000 |
| +5               | 0000 0001       | 0100 0000 |
| +20              | 0000 0100       | 1100 0010 |
| +40              | 0000 1010       | 0000 0000 |
| +80              | 0001 0100       | 0000 0000 |
| +125             | 0001 1111       | 0100 0000 |

#### Table 15. SST Temperature Data Format

#### USING DISCRETE TRANSISTORS

If a discrete transistor is used, the collector is not grounded and should be linked to the base. If a PNP transistor is used, the base is connected to the D1– input and the emitter is connected to the D1+ input. If an NPN transistor is used, the emitter is connected to the D1– input and the base is connected to the D1+ input. Figure 17 shows how to connect the ADT7485A to an NPN or PNP transistor for temperature measurement. To prevent ground noise from interfering with the measurement, the more negative terminal of the sensor is not referenced to ground, but is biased above ground by an internal diode at the D1– input.

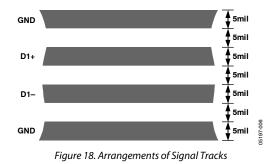




#### LAYOUT CONSIDERATIONS

Digital boards can be electrically noisy environments. Take the following precautions to protect the analog inputs from noise, particularly when measuring the very small voltages from a remote diode sensor:

- Place the ADT7485A as close as possible to the remote sensing diode. Provided that the worst noise sources, such as clock generators, data/address buses, and CRTs, are avoided, this distance can be four to eight inches.
- Route the D1+ and D1- tracks close together in parallel with grounded guard tracks on each side. Provide a ground plane under the tracks if possible.
- Use wide tracks to minimize inductance and reduce noise pickup. A 5 mil track minimum width and spacing is recommended.



- Try to minimize the number of copper/solder joints, which can cause thermocouple effects. Where copper/solder joints are used, make sure that they are in both the D1+ and D1- paths and are at the same temperature.
- Thermocouple effects should not be a major problem because 1°C corresponds to about 240  $\mu$ V, and thermocouple voltages are about 3  $\mu$ V/°C of the temperature difference. Unless there are two thermocouples with a big temperature differential between them, thermocouple voltages should be much less than 200 mV.
- Place a 0.1  $\mu$ F bypass capacitor close to the ADT7485A.
- If the distance to the remote sensor is more than eight inches, the use of a twisted-pair cable is recommended. This works for distances of about 6 to 12 feet.
- For very long distances (up to 100 feet), use shielded twisted-pair cables, such as Belden #8451 microphone cables. Connect the twisted-pair cable to D1+ and D1- and the shield to GND, close to the ADT7485A. Leave the remote end of the shield unconnected to avoid ground loops.

Because the measurement technique uses switched current sources, excessive cable and/or filter capacitance can affect the measurement. When using long cables, the filter capacitor can be reduced or removed. Cable resistance can also introduce errors. A 1  $\Omega$  series resistance introduces about 0.5°C error.

#### **TEMPERATURE OFFSET**

As CPUs run faster, it is more difficult to avoid high frequency clocks when routing the D1+ and D1– tracks around a system board. Even when the recommended layout guidelines are followed, there may still be temperature errors, attributed to noise being coupled on to the D1+ and D1– lines. High frequency noise generally has the effect of producing temperature measurements that are consistently too high by a specific amount. The ADT7485A has a temperature offset command code of 0xe0 through which a desired offset can be set. By doing a one-time calibration of the system, the offset caused by system board noise can be calculated and nulled by specifying it in the ADT7485A. The offset is automatically added to every temperature measurement. The maximum offset is  $\pm 128^{\circ}$ C with 0.25°C resolution. The offset format is the same as the temperature data format—16-bit, twos complement notation, as shown in Table 15. The offset should be programmed in little endian format, that is, LSB before MSB. The offset value is also returned in little endian format when read.

### **OUTLINE DIMENSIONS**

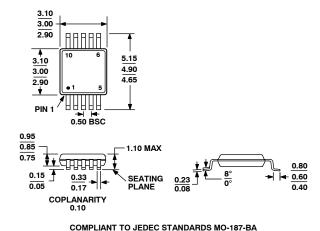


Figure 19. 10-Lead Mini Small Outline Package [MSOP] (RM-10) Dimensions shown in millimeters

#### **ORDERING GUIDE**

| Model                       | Temperature Range | Package Description | Package Option | Branding |
|-----------------------------|-------------------|---------------------|----------------|----------|
| ADT7485AARMZ-R <sup>1</sup> | –40°C to +125°C   | 10-Lead MSOP        | RM-10          | T21      |
| ADT7485AARMZ-R71            | –40°C to +125°C   | 10-Lead MSOP        | RM-10          | T21      |

 $^{1}$  Z = Pb-free part.

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